

Controllable Flux Coupling for the Integration of Flux Qubits

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Abstract

We show a novel method for controlling the coupling of flux-based qubits by means of a superconducting transformer with variable flux transfer function. The device is realized by inserting a small hysteretic dc SQUID with unshunted junctions, working as a Josephson junction with flux-controlled critical current, in parallel to a superconducting transformer; by varying the magnetic flux coupled to the dc-SQUID, the transfer function for the flux coupled to the transformer can be varied. Measurements carried out on a prototype at 4.2 K show a reduction factor of about 30 between the “on” and the “off” states. We discuss the system characteristics and the experimental results.

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Recently, different types of qubits, all based on Josephson junctions, have been experimentally demonstrated. Flux [1, 2], phase [3, 4] and phase-charge [5] qubits have been operated as single devices, while charge state qubits have also been used in an entangled couple, showing quantum-coherent behavior [6] and operation as a conditional gate [7]. In the implementation of a system of entangled qubits, one of the challenges is the realization of a connection between different qubits that fulfills the various constraints imposed by the correct qubit operation. The connection should be non dissipative, otherwise the fluctuations related to its dissipation will destroy the coherent state of the connected qubits; this forbids the use of resistive elements or elements that are dissipative even for a short period of time. It should allow a fast switching, namely its switching time should be much faster than the time related to the clock period. It should be simple and reliable, to be integrable with a large numbers of qubits, and the related implementation should be a well-established technology with a very high degree of reliability. Besides, the coupling strength of the connection should be varied from outside, allowing to turn the coupling on and off whenever needed; it must be noted that a scheme with untunable couplings has been proposed [8] but not yet implemented in practical realizations.

In order to couple flux qubits, it comes natural to use superconducting transformers. Two schemes for achieving coupling control have been presented recently. In the INSQUID [9], which was originally ideated for readout, the flux qubit is placed inside the dc-SQUID of a double-SQUID. The tunable transformer of ref. [10], instead, is conceived for gradiometric flux qubits like that of ref. [1] and is based on the balancing of a gradiometric transformer by means of two small dc-SQUIDs inserted in the transformer branches.

In this letter we propose a Controllable Flux Coupling (CFC) system, suitable for the connection of one or more flux qubits. The CFC basic idea is to use a superconducting flux transformer, modified with the insertion of a small hysteretic dc SQUID that behaves as a Josephson junction with tunable critical current. By modulating the SQUID critical current by means of an external magnetic field, it is possible to control the flux transfer function through the transformer and therefore the coupling. Compared to other proposed schemes, the CFC has the advantage of being easily coupled to a flux qubit through inductively coupled coils, without requiring a particular geometry for the qubit to be read out.

Our variable transformer (Fig. 1) consists of two arms of inductance L , in parallel with a dc-SQUID that performs the control of the magnetic flux transfer. The inner dc-SQUID is a

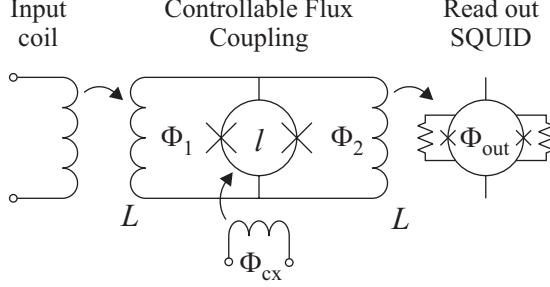


FIG. 1: Schematic of the CFC circuit. The main part is the central gradiometric transformer, with arms of inductance L , modified by an inner dc-SQUID of smaller inductance l ; the coordinates are the magnetic fluxes Φ_1 and Φ_2 . A signal flux is applied on the left side and the CFC response is read out by the SQUID on the right side. The control on the transmitted flux is achieved by coupling a flux Φ_{cx} to the inner dc-SQUID.

loop of inductance $l \ll L$, interrupted by two Josephson junctions of critical current I_0 and capacitance C ; its dynamics is described by the differences of the superconducting phase across the Josephson junctions, which are linked by the fluxoid equation to the magnetic fluxes Φ_1 and Φ_2 in the left and in the right loops.

Input flux is coupled to the left side through an inductively coupled coil, with flux transforming ratio \mathcal{R}_1 ; the flux appearing in the left arm is called Φ_{1x} . The flux response Φ_2 of the variable transformer is read out by a SQUID magnetometer, magnetically coupled to the right side of the transformer with transforming ratio \mathcal{R}_2 ; the measured quantity is then the flux $\Phi_{out} = \mathcal{R}_2\Phi_2$. A control magnetic flux Φ_{cx} linked to the inner dc-SQUID modifies the device behavior. Here and in the following, the subscript x refers to externally applied signals, while c refers to the control loop.

We introduce new coordinates given by linear combinations of the fluxes, $\varphi = \pi(\Phi_2 - \Phi_1)/\Phi_0$ and $\varphi_c = -2\pi(\Phi_1 + \Phi_2)/\Phi_0$, where $\Phi_0 = h/2e$ is the flux quantum; the corresponding reduced driving fluxes are $\varphi_x = -\pi\Phi_{1x}/\Phi_0$ and $\varphi_{cx} = 2\pi\Phi_{cx}/\Phi_0$. With these coordinates, the 2D potential describing the system dynamics can be written as follows:

$$U(\varphi, \varphi_c) = \frac{\Phi_0^2}{4\pi^2 L_0} \left[\frac{1}{2} (\varphi - \varphi_x)^2 + \frac{1}{2} \gamma (\varphi_c - \varphi_{cx})^2 - \beta_0 \cos \frac{\varphi_c}{2} \cos \varphi \right] \quad (1)$$

where $L_0 = L/2 + l/4$ is an effective inductance, $\beta_0 = 2\pi(2I_0)L_0/\Phi_0$ is the corresponding

reduced inductance, with twice the critical current I_0 because of the two junctions in the dc-SQUID, and $\gamma = L_0/l$. Eq.1 has the same form of the potential for a double SQUID [11], with parameters that take into account the gradiometric structure of the device. In the limit for an inner dc-SQUID with a vanishingly small inductance like in our case, i.e. $\gamma \gg 1$, the degree of freedom related to φ_c is frozen and restrained to an equilibrium value $\varphi_c \simeq \varphi_{cx}$, so that the potential becomes a 1D curve in the remaining coordinate φ :

$$U(\varphi) = E_J \left[\frac{1}{2\beta} (\varphi - \varphi_x)^2 - \cos \varphi \right] \quad (2)$$

where $E_J = (2I_0)\Phi_0/(2\pi)\cos(\pi\Phi_{cx}/\Phi_0)$ and $\beta = \beta_0\cos(\pi\Phi_{cx}/\Phi_0)$. This is equivalent to the potential of an ordinary rf-SQUID, but here the critical current can be modulated by the external flux Φ_{cx} linked to the loop of the inner dc-SQUID and the role of reduced inductance is played by the quantity β that is not restrained to assume only positive values. By deriving eq. 2 with respect to φ and setting the derivative to zero, we find the relationship linking φ and the excitation φ_x to find the extremal points:

$$\varphi_x = \varphi + [\beta_0\cos(\pi\Phi_{cx}/\Phi_0)]\sin\varphi \quad (3)$$

For $|\beta| \leq 1$ the relation is single-valued and the potential presents just one minimum, while for $|\beta| > 1$ the relation is multi-valued, with more minima separated by potential barriers.

An input signal centered around $\varphi_x = 0$, with amplitude smaller than a flux quantum, causes a monotonic and single-valued flux response φ , whose amplitude depends on the control parameter φ_{cx} . In a sufficiently small region this response is linear and hence the system behaves as a linear controllable transformer. In this regime the overall transfer parameter \mathcal{R} , namely which part of the input magnetic flux is transmitted to the output, is given by the slope of the flux characteristics at the flex point $\varphi_x = 0$. By returning to the quantities Φ_{1x} (flux coupled to the left arm of the transformer), Φ_2 (transformer response flux) and Φ_{cx} (flux coupled to the inner dc-SQUID), one can write:

$$\mathcal{R} = \frac{d\Phi_2}{d\Phi_{1x}}|_{\Phi_{1x}=0} = \frac{1}{|1 + \beta_0\cos\frac{\pi\Phi_{cx}}{\Phi_0}|} \quad (4)$$

This modulation of the overall transfer parameter \mathcal{R} , achieved acting on the flux Φ_{cx} , is the feature that we exploit to obtain a tunable transformer: while keeping the flux working point around zero, the potential is changed in such a way to change the shape of the char-

acteristics and operate between two points with very different responsivity (the “on” and the “off” states). During operation, the system is kept in the same potential well, avoiding sudden dissipative jumps of the system to other minima; besides, the potential change must be fast but still adiabatic. This last requirement represents the main limit on the operating speed of the CFC system; for our test device one can estimate this limit from the plasma frequency $f_p = 1/2\pi\sqrt{LC} \sim 10GHz$, obtaining a value suitable for typical superconducting quantum gates operations.

In order to test the features of the variable transformer, we built an integrated device composed of transformer, excitation coil and non-hysteretic readout dc-SQUID, using trilayer Nb/AlOx/Nb technology. The inner dc-SQUID [12] is made by two loops in a gradiometric configuration, with an area of $10\mu m \times 10\mu m$, partially covered by the overlaying Nb layer; the total inductance has been evaluated in previous measurements to be about $5pH$. The junctions have nominally $3\mu m$ side and a critical current $I_0 \simeq 5\mu A$, measured in a similar, isolated device; the corresponding reduced inductance β_l , then, is much less than unity. The inner dc-SQUID is connected in parallel to two Nb coils of inductance L , each consisting of two turns wound around a square of $100\mu m$ side (computed value $L = 628pH$). The transformer arms are magnetically coupled respectively to the excitation coil and to the input coil of the readout dc-SQUID, both made of two turns, nested into the rf-SQUID loops and tightly coupled by means of a ground plane; the measured transfer ratio to the readout SQUID is $\mathcal{R}_2 = 0.20$. For this variable transformer, the maximum value of the reduced inductance β_0 is 9.5.

In the experiment, carried out at $4.2K$, we measured the flux response to a sweeping external flux for different values of the flux Φ_{cx} applied to the inner dc-SQUID. The experimental curves are shown in Fig. 2a; for clarity, curves with hysteretic behavior, which have been observed in the device, are not displayed. The curves are described by Eq. 3, except for a small shift, both in the horizontal and vertical directions, caused by the unavoidable spurious coupling of the control flux Φ_{cx} to the transformer input coil and to the readout SQUID. From the measured shifts we estimated the values of the mutual inductance between the control flux coil and the transformer input ($M_{c,1} = 2.6pH$), and that between the control flux coil and the readout SQUID ($M_{c,out} = 0.29pH$). Fig. 2b shows the experimental curves after correction for the spurious coupling. It was verified experimentally that the input flux Φ_{1x} produces a negligible spurious effect both on the readout SQUID and on the

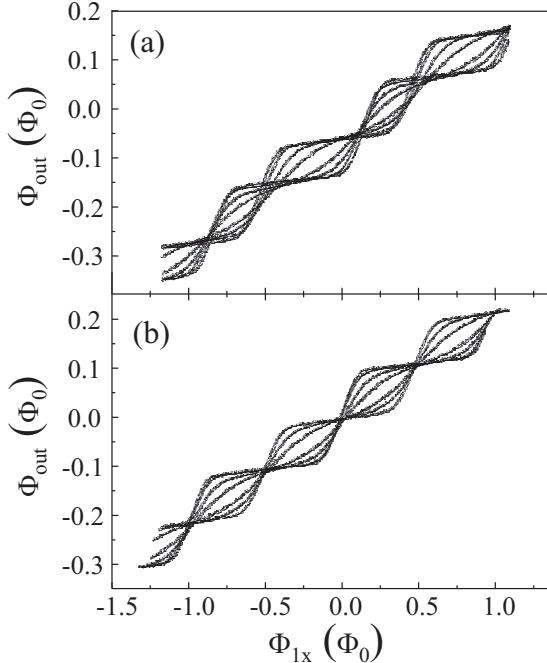


FIG. 2: a) Experimental curves showing the flux characteristics of the device for different values of the control magnetic flux Φ_{cx} , all in the non-hysteretic region. b) Data corrected by removing the effect of spurious couplings.

inner dc-SQUID.

Fig. 3 shows the measured slopes of the acquired characteristics at the working point, namely at the crossing closest to $\Phi_{1x} = 0$ in Fig. 2a, as a function of the control flux. The fit (continuous line) is made using Eq. 4, with spurious coupling taken into account, and allows an independent estimate of β_0 that is consistent with the project value. While the agreement is very good in the bottom part of the curve, in the upper part the experimental points are lower than expected. This effect is due to the rounding of the flux characteristics because of thermal fluctuations: the slope in the steepest points is reduced. At lower temperature this smearing effect is expected to decrease with the square root of the temperature, until the classical-quantum crossover temperature is reached (hundreds of mK for our devices) and quantum tunnelling becomes the dominant fluctuation term. The ratio between minimum and maximum slope (about 30 for the data of Fig. 3) is a figure of merit for the performance of the variable transformer.

Let's now discuss how the variable transformer works. The “off” state, where the transfer ratio is minimum, is easily identified with the state where the effective reduced inductance

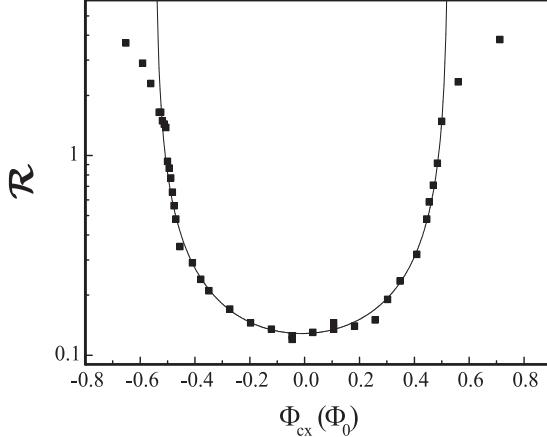


FIG. 3: Experimental values of the slope of the flux characteristics at the working point (zero input flux), plotted against the value of the control flux, coupled to the inner dc-SQUID. The continuous line is the best fit with Eq. 4, considering also the spurious direct coupling between input and output.

is maximum, that is $\mathcal{R}_{min} = 1/(1 + \beta_0)$; to get a small flux transfer, then, the device must be highly hysteretic (large β_0). In this condition the slope of the flux characteristic is almost horizontal and a large amount of input flux Φ_{1x} can be fed while keeping the device in the same flux state; this curve is not shown in Fig. 2. By increasing the control flux to $\Phi_0/2$, one gets $\mathcal{R} \simeq 1$, and the input flux is totally transmitted to the output; however this is not the steepest possible slope, since by further increasing Φ_{cx} one gets β tending to -1 and a diverging \mathcal{R} . Excluding the non-physical point $\beta = -1$, we can then increase the transfer ratio \mathcal{R} beyond unity. Two considerations must be done at this point. First, while the “off” state can be obtained with a hysteretic characteristic (provided that there are no transitions between different flux states), the “on” state must be obtained with non-hysteretic characteristics and this restrains the range of usable values to $|\beta| < 1$. Second, the flux response is not necessarily linear with the input flux; the dynamical range where linearity is ensured is depending on the parameter β : the steepest is the slope of the flux characteristic, the smaller is the allowed flux range. With larger flux signals, the response has a saturated amplitude and harmonics are produced. However, in certain cases this may not be a limitation: in qubit operation, in fact, generally one has just to distinguish between the two different flux states, the required response being just an identification of the qubit state. In this situation, operating in the non linear range does not affect the efficiency of the

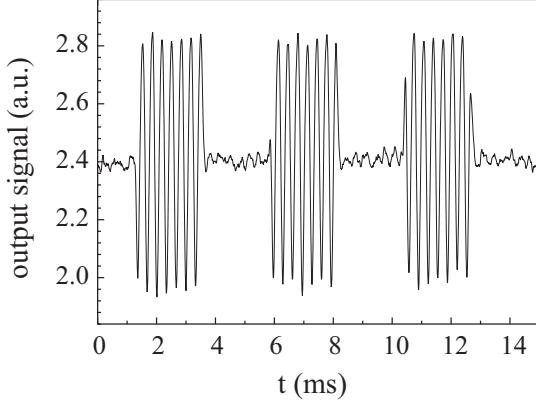


FIG. 4: Modulation of a sinusoidal signal fed on one side of the tunable transformer. The ratio of the transmitted amplitude between on and off states is 30.

measurement and extends the usable range of the input signals. As a matter of fact, choosing the working points for “off” and “on” states is best achieved experimentally, according to the specific requirements of the experiment and to the signal characteristics that must be preserved.

To test the performance of the transformer, we sent a sinusoidal signal to the left side of the transformer and measured the output from the readout dc-SQUID while modulating the inner dc-SQUID with a square wave between the points of maximum and minimum transfer ratio, chosen experimentally.

The resulting modulation is shown in Fig. 4. The measured efficiency, the maximum ratio between the “on” and the “off” transmission, is about 30, in agreement with the data of Fig. 4. For this situation, the measured range in which the response is linear corresponds a peak-to-peak input flux of $0.1\Phi_0$. If the linear range is exceeded, the efficiency in the flux modulation decreases but operation is still possible. As an example, for an input flux of $0.3\Phi_0$ peak-to-peak the measured ratio between the “on” and the “off” transmission is about 14.

CFC systems can be integrated together with the SQUID flux qubits, since they are based on the same technology, and can be used to control the couplings between them. They can also be used to realize a bus for the controlled coupling of more qubits, for example by using the scheme shown in Fig. 5; in this example a pair of qubits can be coupled by switching “on” the relative “switches”, and by maintaining all the others in the “off” state (this scheme is similar to that proposed in [13]). Since the CFC remains always in the

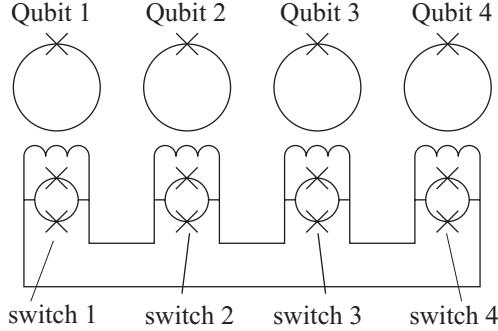


FIG. 5: Example of a possible scheme for the controlled coupling of more flux qubits.

superconducting state without jumps to a dissipative state, the only contribution to the overall decoherence is due to the device intrinsic dissipation and to the environmental noise pick-up. This means that the total contribution to decoherence should be of the same order of the qubit contribution, since they are very similar for technology, dimensions, components and structure.

In conclusion, we have realized and tested a microfabricated SQUID based controllable flux coupling, useful in any application in which it is necessary to modify the magnetic coupling between different devices, and in particular suitable for quantum computing applications with flux qubits.

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